

Frequency domain analysis (of the GPR transmitted field)

Silvestar Šesnić

Faculty of Electrical Engineering, Mechanical Engineering
and Naval Architecture
University of Split, Croatia

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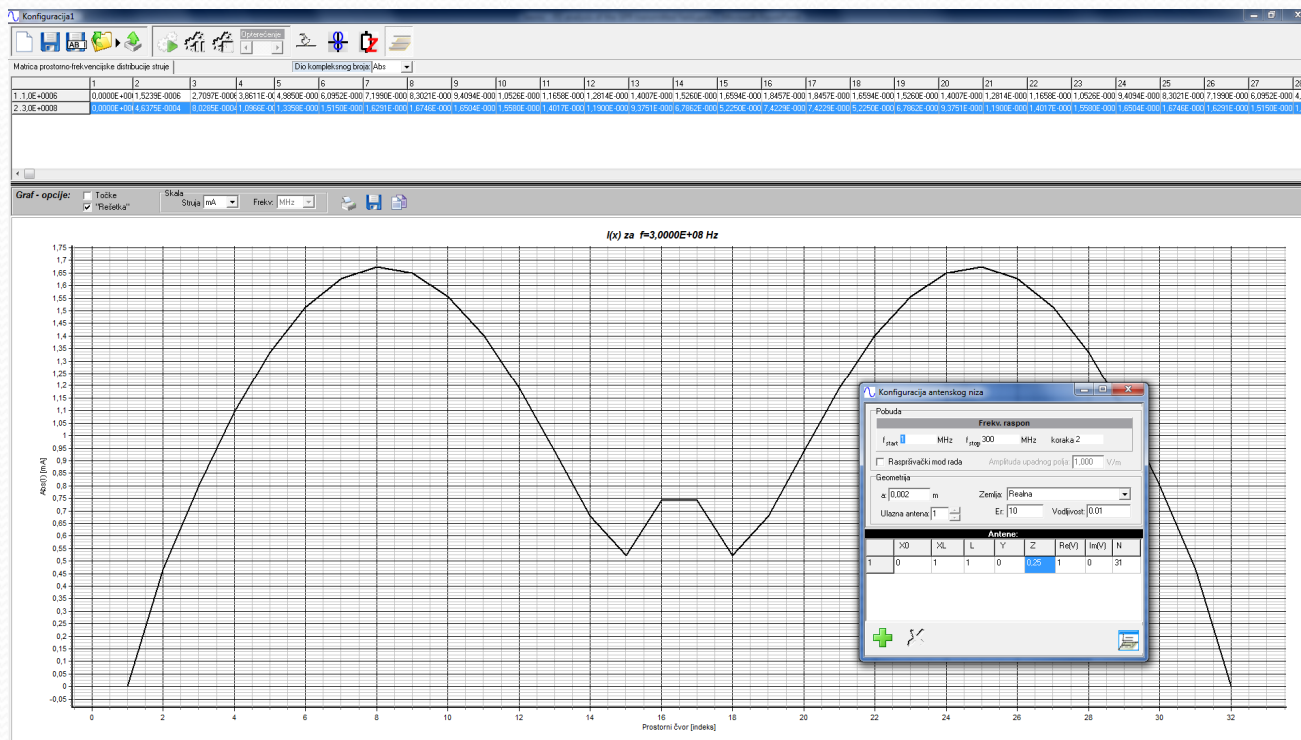
- Introduction to software SuzanaFD
- Analytical solutions for the wire antenna in the FD
- Transmitted electrical field in the FD

SuzanaFD

Introduction

SuzanaFD

- Introduction to software SuzanaFD



Integral equations

...and methods for their solution

Pocklington equation (1897.)

Frequency domain

$$E_x^{exc}(\omega) = -\frac{1}{j4\pi\omega\epsilon_{eff}} \int_0^L \left(\frac{\partial^2}{\partial x^2} - \gamma^2 \right) I(x', \omega) g(x, x') dx'$$

Time domain

$$\left(\epsilon \frac{\partial}{\partial t} + \sigma \right) E_x^{exc}(t) = - \left(\frac{\partial^2}{\partial x^2} - \mu\sigma \frac{\partial}{\partial t} - \mu\epsilon \frac{\partial^2}{\partial t^2} \right) \int_0^L \frac{I\left(x', t - \frac{R}{v}\right)}{4\pi R} dx'$$

Hallen equation (1938.)

Frequency domain

$$\int_0^L I(x', \omega) g(x, x') dx' = C \cos(kx) + B \sin(kx) - \frac{j4\pi}{Z_0} \int_0^L E_x^{exc}(x', \omega) \sin[k(x - x')] dx'$$

Time domain

$$\int_0^L \frac{I\left(x', t - \frac{R}{c}\right)}{4\pi R} dx' = F_0\left(t - \frac{x}{c}\right) + F_L\left(t - \frac{L - x}{c}\right) + \frac{1}{2Z_0} \int_0^L E_x^{exc}\left(x', t - \frac{|x - x'|}{c}\right) dx'$$

Methods of solving integral equations

- Numerical solutions
 - Complex geometries
 - Wide range of parameters
- Analytical solutions
 - Canonical geometries
 - Limited parameters
 - Use of approximations
 - Procedure control
 - Simple solutions

Analytical solution

Frequency domain approximations and procedure

Frequency domain approximations – thin-wire approximation

$$\vec{A}(\vec{r}, \omega) = \frac{\mu}{4\pi} \int_{V'} \vec{J}(\vec{r}') g(\vec{r}, \vec{r}') dV' \quad A_x(x, \omega) = \frac{\mu}{4\pi} \int_0^L I(x') g(x, x') dx'$$

$$\vec{E} = -\nabla \varphi - j\omega \vec{A}$$

$$E_x = -\frac{\partial \varphi}{\partial x} - j\omega A_x$$

Frequency domain approximations – reflection coefficient

$$U_{11} = 2 \int_0^{\infty} \frac{e^{-2\mu_1(z+h)}}{\mu_1 + \mu_2} J_0(\lambda\rho) \lambda d\lambda$$

$$W_{11} = 2 \int_0^{\infty} \frac{(\mu_1 - \mu_2) e^{-2\mu_1(z+h)}}{k_2^2 \mu_1 + k_1^2 \mu_2} J_0(\lambda\rho) \lambda d\lambda$$

Sommerfeld integrals

$$\Gamma_{ref} = \frac{\frac{1}{\epsilon_r + \frac{\sigma}{j\omega\epsilon_0}} \cos \theta - \sqrt{\frac{1}{\epsilon_r + \frac{\sigma}{j\omega\epsilon_0}} - \sin^2 \theta}}{\frac{1}{\epsilon_r + \frac{\sigma}{j\omega\epsilon_0}} \cos \theta + \sqrt{\frac{1}{\epsilon_r + \frac{\sigma}{j\omega\epsilon_0}} - \sin^2 \theta}}$$

Fresnel reflection coefficient approach

Frequency domain approximations – integral function

$$\int_0^L I(x') g(x, x') dx' = \int_0^L I(x) g(x, x') dx' + \int_0^L [I(x') - I(x)] g(x, x') dx'$$

$$I(x') - I(x) \equiv 0$$

$$\int_0^L I(x') g(x, x') dx' = I(x) \int_0^L g(x, x') dx'$$

Green's function

$$\int_0^L g(x, x') dx' = \begin{cases} 2 \ln \frac{L}{a} \\ 2 \left(\ln \frac{L}{a} - \Gamma_0^{ref} \ln \frac{L}{2h} \right) \\ 2 \left(\ln \frac{L}{a} - \Gamma_E^{ref} \ln \frac{L}{2d} \right) \end{cases}$$

Free space

Above lossy ground

Below lossy ground

$$\frac{\partial^2 I(x, \omega)}{\partial x^2} - \gamma^2 I(x, \omega) = j \frac{4\pi\gamma^2}{\omega\mu\Psi(\omega)} E_x(\omega)$$

Differential equation

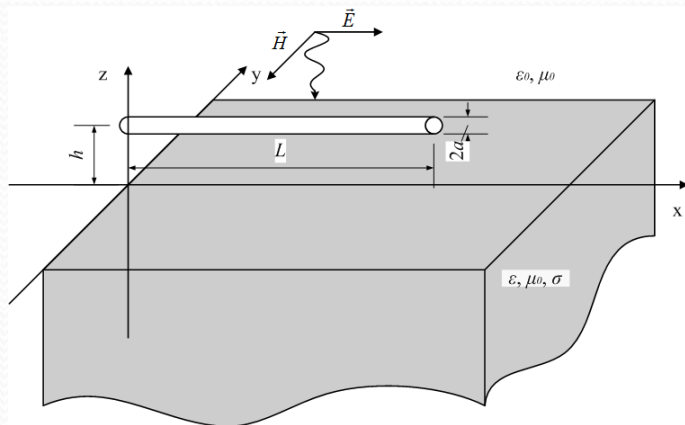
Induced current

$$I(x, \omega) = \frac{4\pi}{j\omega\mu\Psi(\omega)} E_x(\omega) \left[1 - \frac{\cosh\left(\gamma\left(\frac{L}{2} - x\right)\right)}{\cosh\left(\gamma\frac{L}{2}\right)} \right]$$

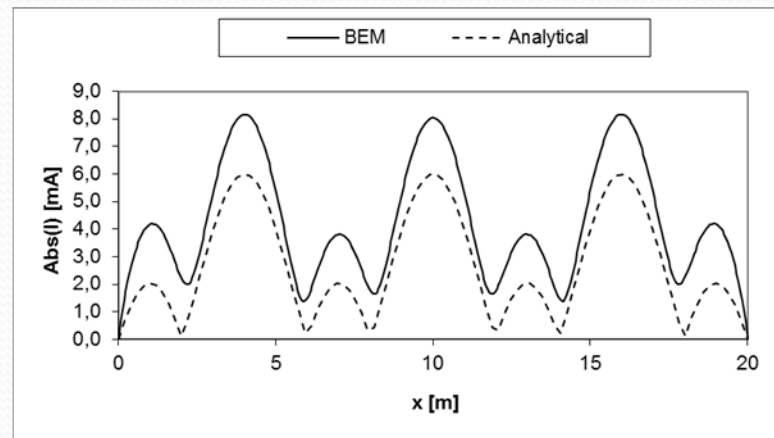
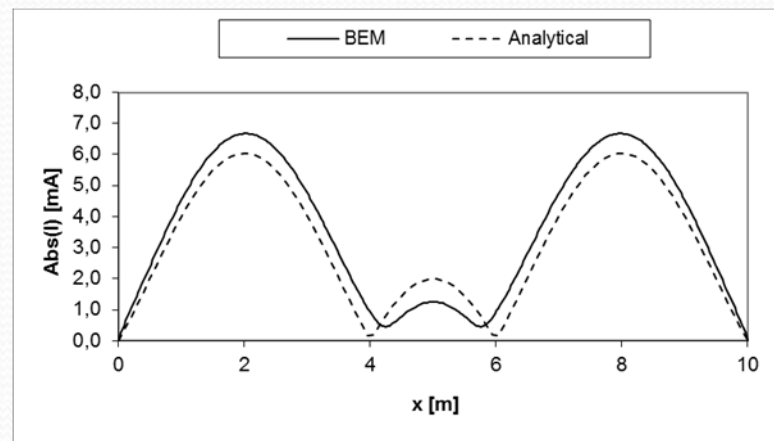
Analytical solutions

Examples

Wire above ground (power cable...)

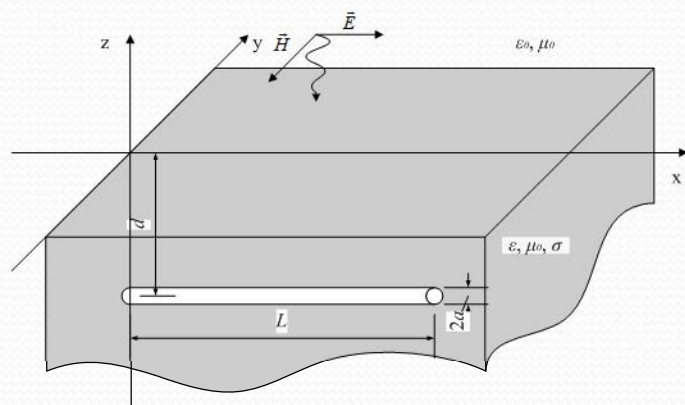


$$I(x, \omega) = \frac{4\pi}{j\omega\mu\Psi(\omega)} E_x^{exc}(\omega) \left[1 - \frac{\cos\left(k\left(\frac{L}{2} - x\right)\right)}{\cos\left(k\frac{L}{2}\right)} \right]$$

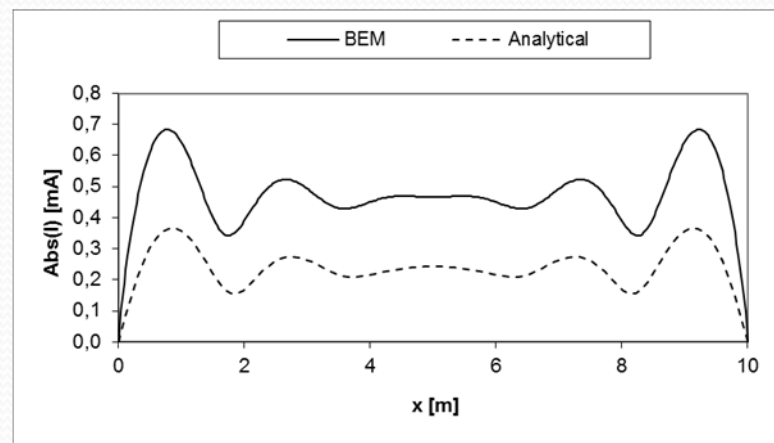
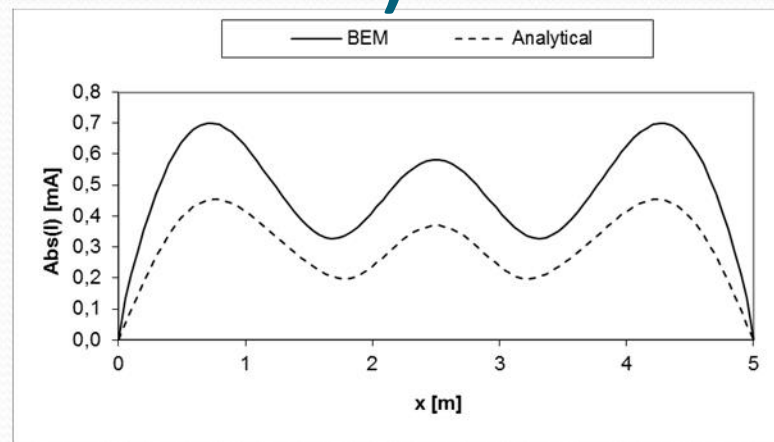


$f=50$ MHz

Wire below ground (telecommunication cable...)



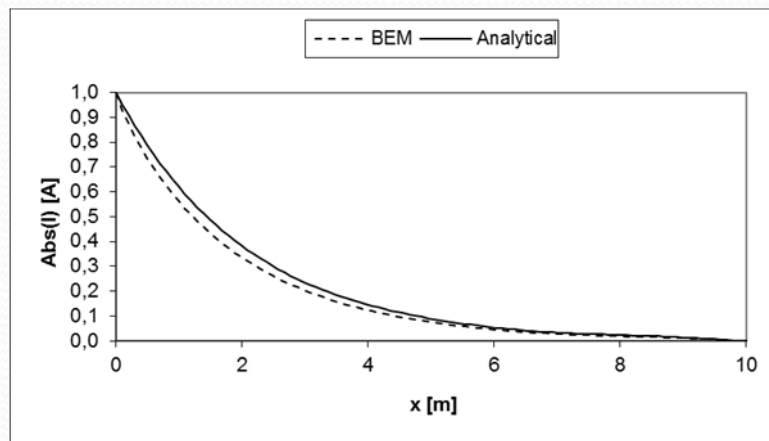
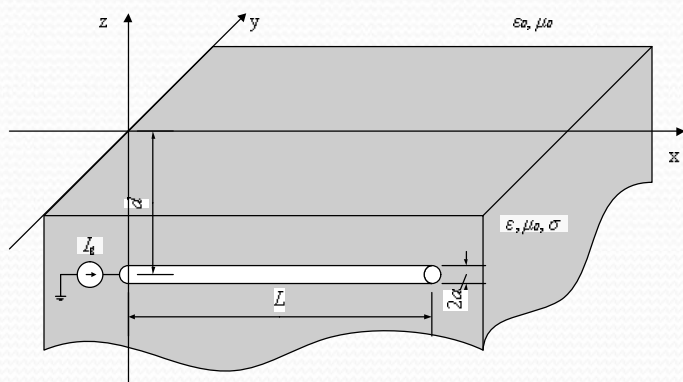
$$I(x, \omega) = \frac{4\pi e^{j\frac{a}{v}\omega}}{j\omega\mu\Psi(\omega)} E_x^{exc}(\omega) \left[1 - \frac{\cosh\left(\gamma\left(\frac{L}{2} - x\right)\right)}{\cosh\left(\gamma\frac{L}{2}\right)} \right]$$



$f=50$ MHz

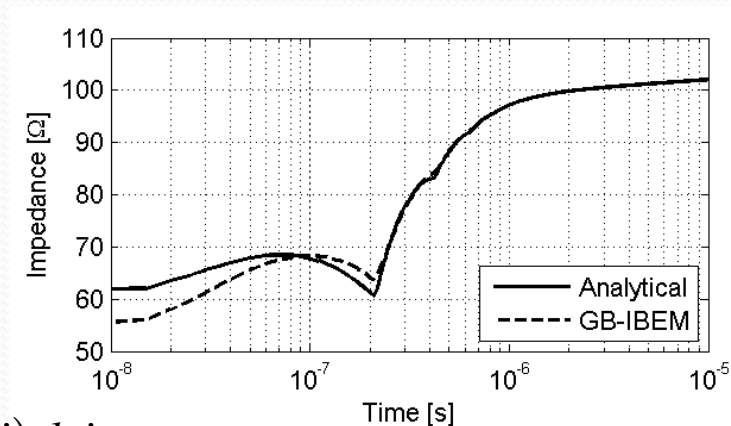
Wire below ground (grounding electrode...)

$f=10$ MHz

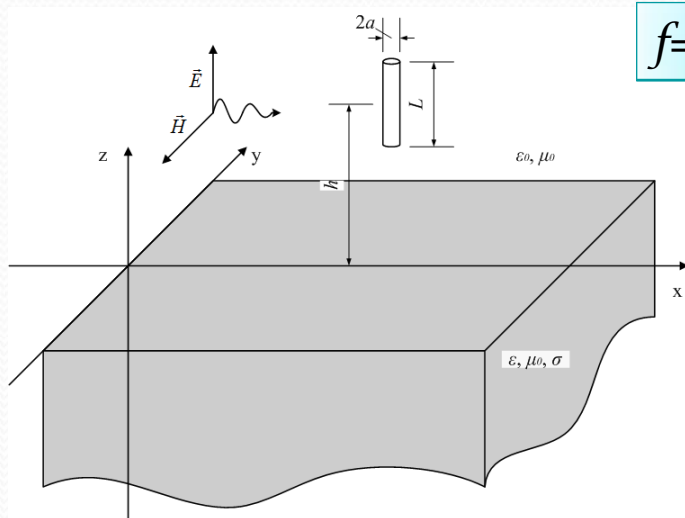


$$I(x, \omega) = I_g(\omega) \frac{\sinh[\gamma(L-x)]}{\sinh(\gamma L)}$$

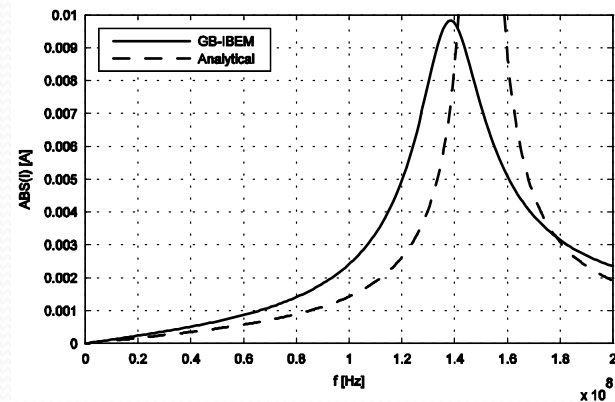
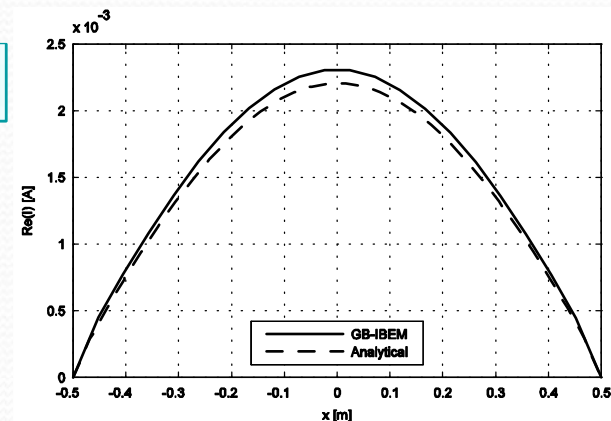
$$V^{sct}(x, \omega) = \frac{\gamma I_g(\omega)}{j4\pi\omega\epsilon_{eff} \sinh(\gamma L)} \int_0^L \cosh[\gamma(L-x)] g(x, x') dx'$$



Vertical wire above ground (receiving antenna...)



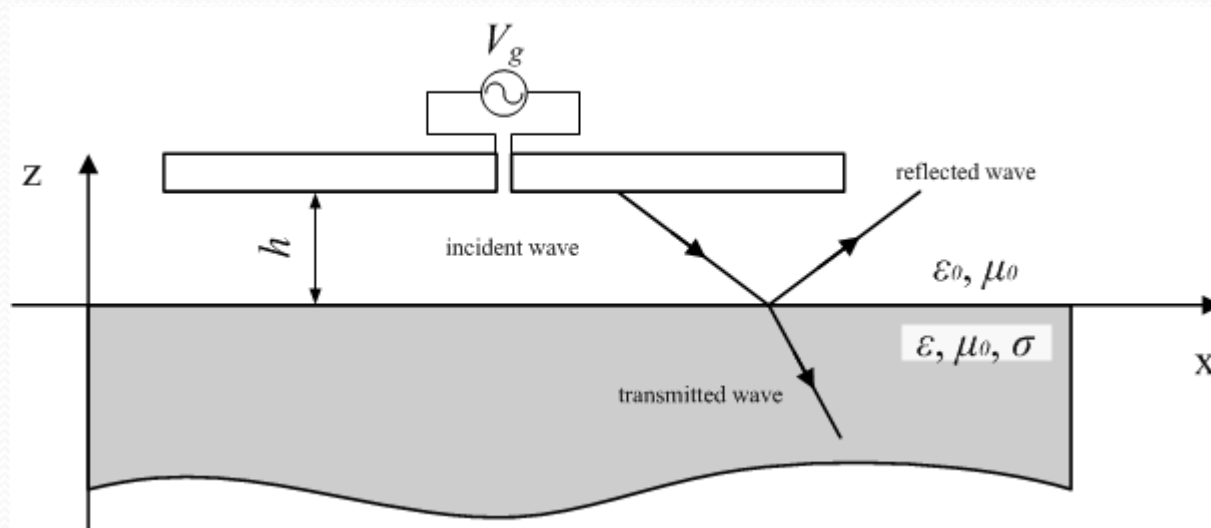
$f=168.2 \text{ MHz}$



$$I(z, \omega) = \frac{4\pi}{j\omega\mu\Psi(z, \omega)} E_z^{exc}(\omega) \left[1 - \frac{\cos(k(h-z))}{\cos\left(k\frac{L}{2}\right)} \right]$$

Transmitted field - formulation

Transmitting GPR dipole antenna



Pocklington integral equation

$$E_x^{exc}(\omega) = j\omega \frac{\mu}{4\pi} \int_0^L I(x', \omega) g(x, x') dx'$$
$$- \frac{1}{j4\pi\omega\epsilon} \frac{\partial}{\partial x} \int_0^L \frac{\partial I(x', \omega)}{\partial x'} g(x, x') dx'$$

Transmitted field in the lower medium

$$E_x^{tr}(x, z, \omega) = -\frac{1}{j4\pi\omega\epsilon_{eff}} \left[\int_0^L \frac{\partial I(x', \omega)}{\partial x'} \frac{\partial G(x, x', z)}{\partial x'} dx' + \gamma^2 \int_0^L I(x', \omega) G(x, x', z) dx' \right]$$

$$E_z^{tr}(x, z, \omega) = \frac{1}{j4\pi\omega\epsilon_{eff}} \int_0^L \frac{\partial I(x', \omega)}{\partial x'} \frac{\partial G(x, x', z)}{\partial z} dx'$$

Transmitted field in the lower medium

$$G(x, x', z) = \Gamma_{tr}^{MIT} g_E(x, x', z)$$

Green's function

$$\Gamma_{tr}^{MIT} = \frac{2\varepsilon_{eff}}{\varepsilon_{eff} + \varepsilon_0} \quad \varepsilon_{eff} = \varepsilon_r \varepsilon_0 - j \frac{\sigma}{\omega}$$

$$g_E(x, x', z) = \frac{e^{-\gamma R}}{R} \quad \gamma = j\omega \sqrt{\mu \varepsilon_{eff}}$$

Boundary element formalism

$$I(x') = I_{1i} \frac{x_{2i} - x'}{\Delta x} + I_{2i} \frac{x' - x_{1i}}{\Delta x}$$

$$\frac{\partial I(x')}{\partial x'} = \frac{I_{2i} - I_{1i}}{\Delta x}$$

Induced current

Boundary element formalism

$$\begin{aligned}
 E_x^{tr}(x, z, \omega) &= -\frac{1}{j4\pi\omega\epsilon_{eff}} \sum_{i=1}^{N_j} \left[\frac{I_{2i} - I_{1i}}{\Delta x} \int_{x_{1i}}^{x_{2i}} \frac{\partial G(x, x', z)}{\partial x'} dx' \right. \\
 &\quad \left. + \gamma^2 \int_{x_{1i}}^{x_{2i}} \left[I_{1i} \frac{x_{2i} - x'}{\Delta x} + I_{2i} \frac{x' - x_{1i}}{\Delta x} \right] G(x, x', z) dx' \right] \\
 E_z^{tr}(x, z, \omega) &= \frac{1}{j4\pi\omega\epsilon_{eff}} \sum_{i=1}^{N_j} \frac{I_{2i} - I_{1i}}{\Delta x} \int_{x_{1i}}^{x_{2i}} \frac{\partial G(x, x', z)}{\partial z} dx'
 \end{aligned}$$

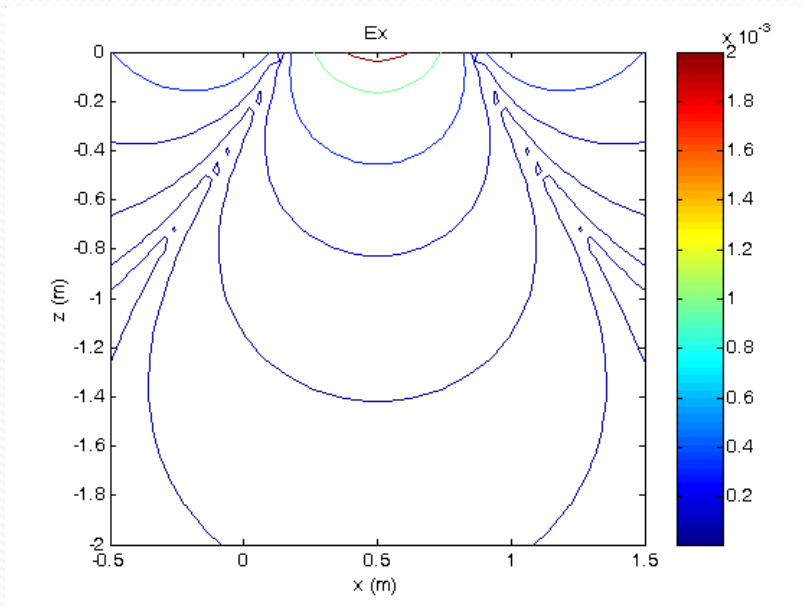
Numerical results

Parameters of the calculation

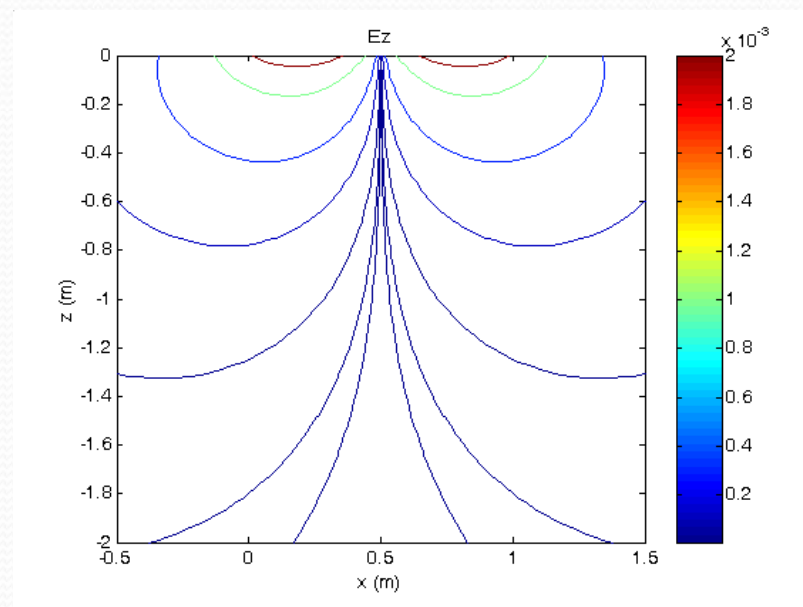
- Properties of the antenna
 - Length $L=1$ m
 - Radius $a=2$ mm
 - Height $h=0.25$ m
- Properties of the lower medium
 - Relative permittivity $\epsilon_r=10$
 - Conductivity $\sigma=10$ mS/m
- Properties of the excitation
 - $V_o=1$ V, $f=1, 10, 100, 300$ MHz

Numerical results, $f=1$ MHz

E_x

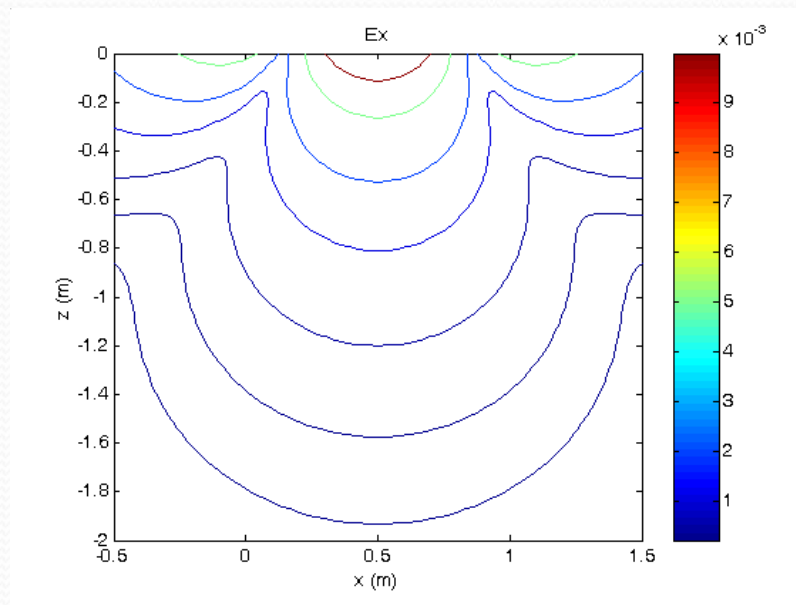


E_z

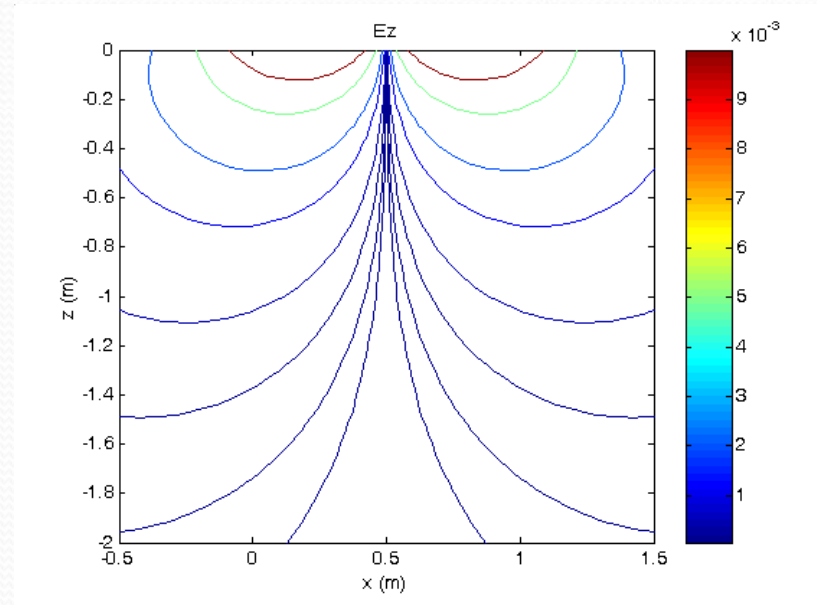


Numerical results, $f=10$ MHz

E_x

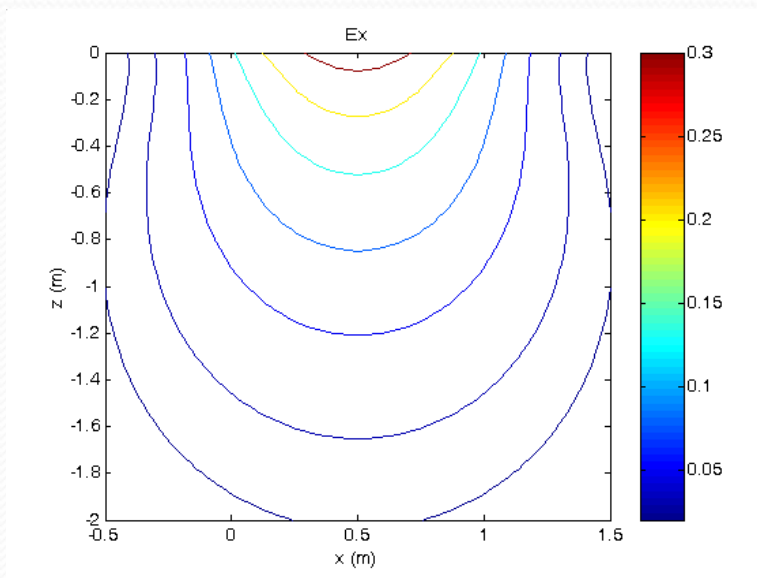


E_z

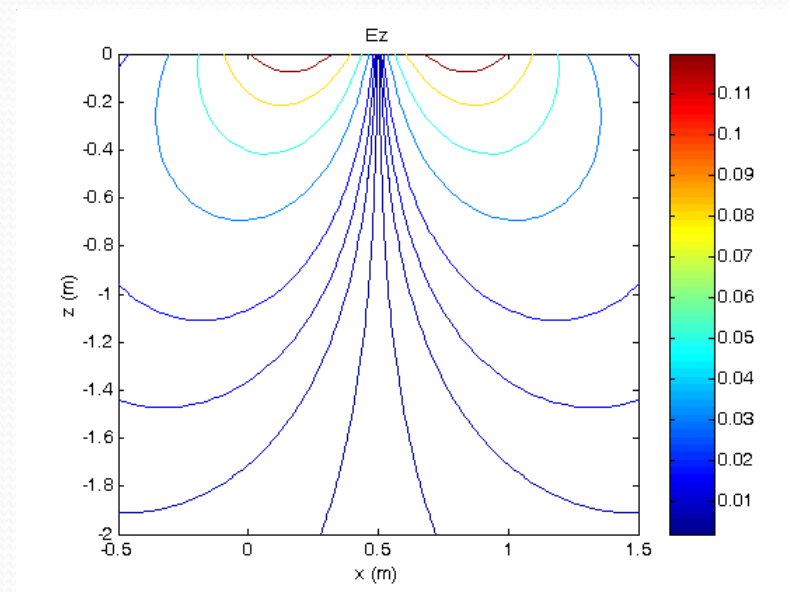


Numerical results, $f=100$ MHz

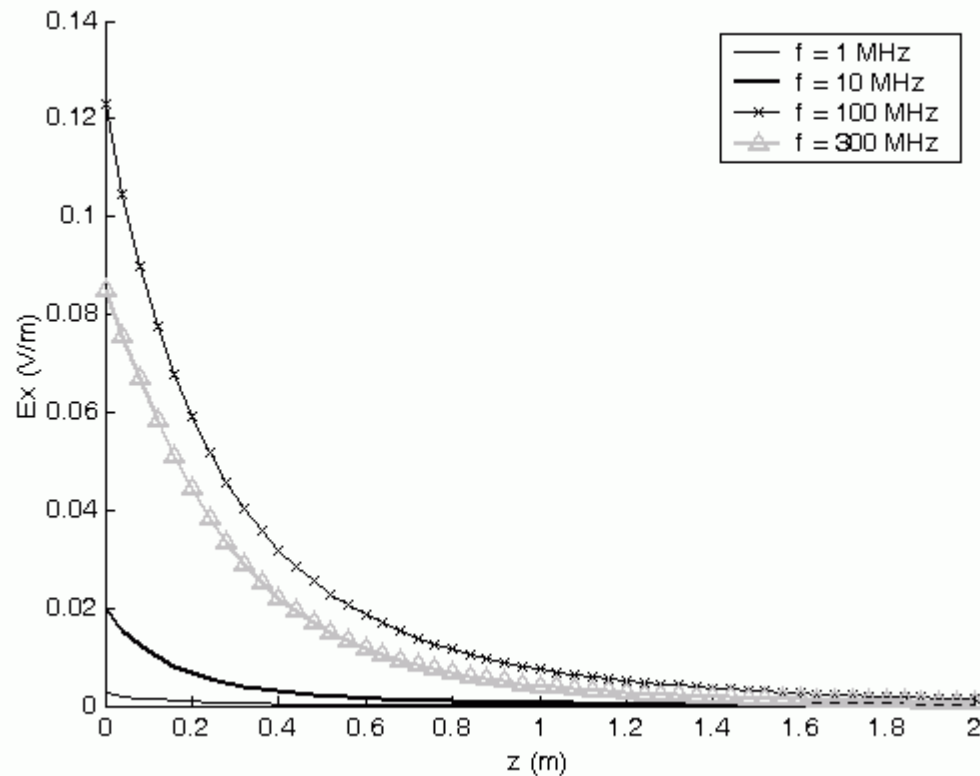
E_x



E_z



Numerical results, broadside field



Concluding remarks

Concluding remarks

- Software SuzanaFD, freely available
- Analytical solutions for the wire antenna in the FD
 - Future work – solution for the transmitting antenna
- Transmitted electrical field in the FD

**Thank you for your
attention!**